

Lecture 32 - The "Short" Metal-Oxide-Semiconductor Field-Effect Transistor (*cont.*)

April 27, 2007

Contents:

1. MOSFET scaling

Reading assignment:

P. K. Ko, "*Approaches to Scaling.*"

Key questions

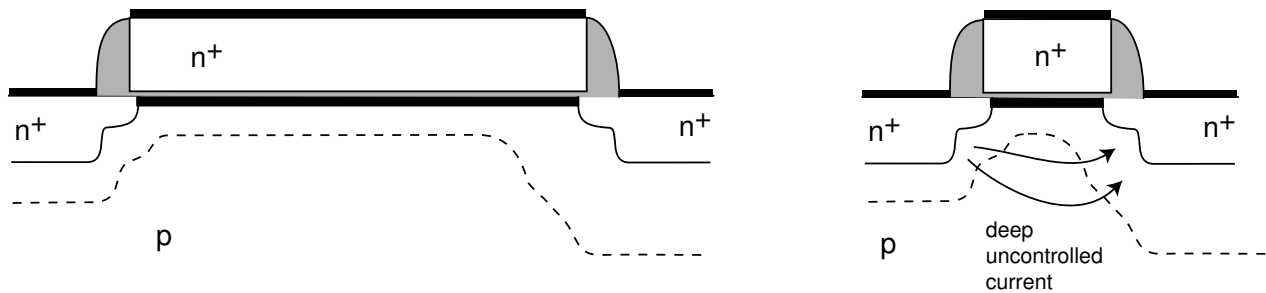
- What happens if a MOSFET gate length is simply shrunk in size without changing anything else?
- How should the MOSFET design change as it shrinks down in size?

1. MOSFET scaling

Several driving forces for scaling down size of MOSFET:

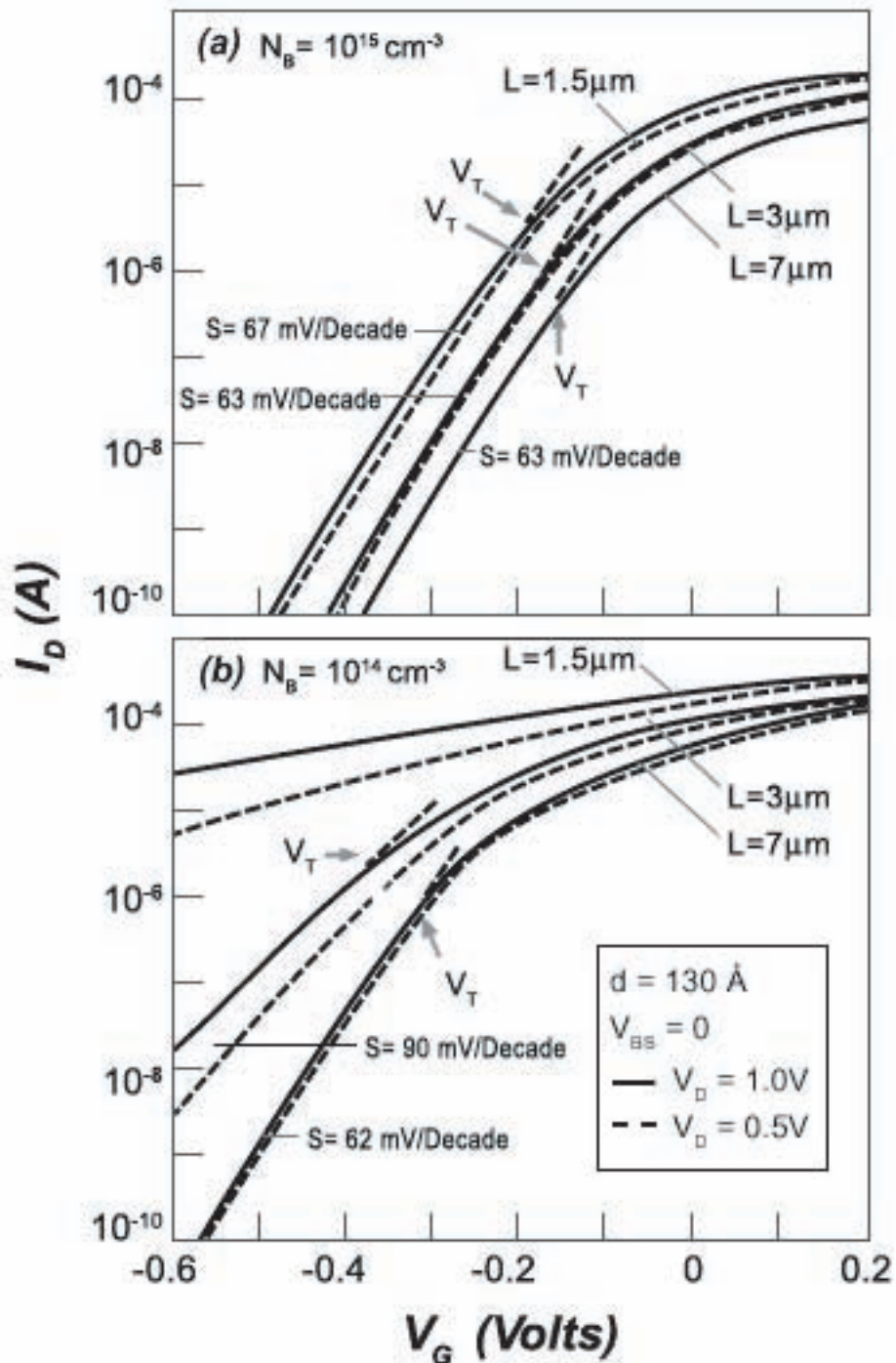
- higher density circuits: SSI, MSI, LSI, VLSI, ULSI, RLSI, ...
- higher performance: $L \downarrow \Rightarrow I_D \uparrow \Rightarrow \tau_{switch} \downarrow$
- lower power consumption: $L \downarrow \Rightarrow V_{DD} \downarrow$

Simple L scaling compromises *electrostatic integrity* and produces *punchthrough* (extreme case of short-channel effects):



To avoid punchthrough:

- $N_A \uparrow \Rightarrow V_T \uparrow \Rightarrow I_D \downarrow$
- $V_{DD} \downarrow \Rightarrow I_D \downarrow$
- $x_{ox} \downarrow \Rightarrow V_T \downarrow \Rightarrow I_D \uparrow$



Subthreshold characteristics for various channel lengths.

(a) $N_A = 10^{15} \text{ cm}^{-3}$, (b) $N_A = 10^{14} \text{ cm}^{-3}$.

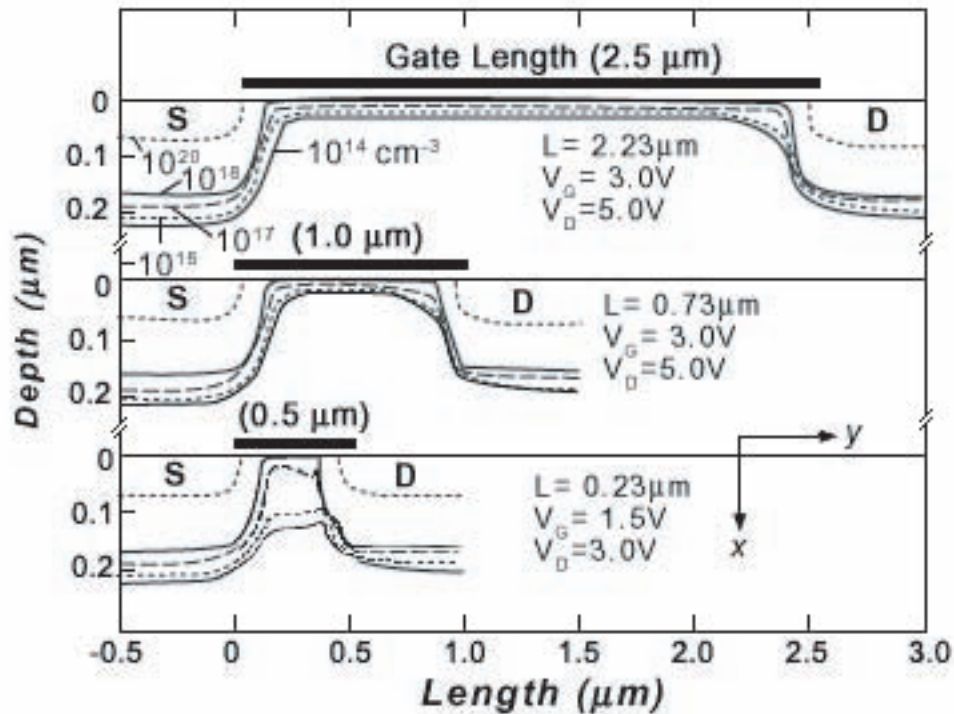
Adapted from S. M. Sze,

Physics of Semiconductor Devices, 2nd ed., Wiley, 1981 (470)

Image by MIT OpenCourseWare.

Adapted from Sze, S. M. *Physics of Semiconductor Devices*. 2nd ed.

New York, NY: John Wiley & Sons, 1981, p. 470. ISBN: 9780471056614.



Constant electron density contours for 3 MOSFETs with channel lengths 2.23, 0.73, and 0.23 μm .

Adapted from S. M. Sze,
Physics of Semiconductor Devices, 2nd ed., Wiley, 1981 (481).

Image by MIT OpenCourseWare.

Adapted from Sze, S. M. *Physics of Semiconductor Devices*. 2nd ed.

New York, NY: John Wiley & Sons, 1981, p. 481. ISBN: 9780471056614.

Need smart way of scaling:

- constant field scaling
- constant voltage scaling
- generalized scaling

□ Constant field scaling

Scale keeping vertical and horizontal electric fields constant.

Define: *scaling factor* $S > 1$

parameter	scaling factor
device dimensions (L, W, x_{ox})	$1/S$
doping level (N_A)	S
supply voltage (V_{DD})	$1/S$

Consequences (use simple long-channel theory):

- gate capacitance:

$$C'_{gs} = C'_{ox} L' W' = S C_{ox} \frac{L W}{S S} = \frac{C_{gs}}{S} \downarrow$$

- threshold voltage:

$$V'_T = V_{FB} + \phi_{sth} + \gamma \sqrt{\phi_{sth}} \simeq \frac{1}{C'_{ox}} \sqrt{2\epsilon_s q N'_A \phi_{sth}} \sim \frac{V_T}{\sqrt{S}} \downarrow$$

- drive current:

$$I'_D = \frac{W'}{2L'} \mu_e C'_{ox} (V'_{DD} - V'_T)^2 = \frac{W}{2L} \mu_e S C_{ox} \left(\frac{V_{DD}}{S} - \frac{V_T}{\sqrt{S}} \right)^2 = \frac{I_D}{S} \downarrow$$

- gate delay:

$$\tau' = \frac{C'_{gs} V'_{DD}}{I'_D} = \frac{\frac{C_{gs}}{S} \frac{V_{DD}}{S}}{\frac{I_D}{S}} = \frac{\tau}{S} \downarrow$$

- power-delay product or *switching energy*:

$$C'_{gs} V'_{DD}{}^2 = \frac{C_{gs}}{S} \left(\frac{V_{DD}}{S} \right)^2 = \frac{C_{gs} V_{DD}^2}{S^3} \downarrow\downarrow\downarrow$$

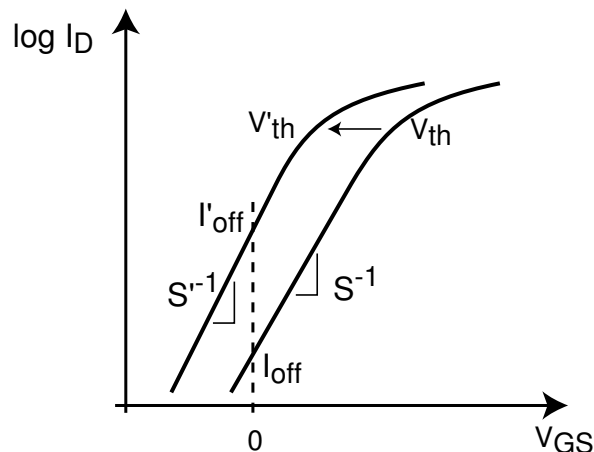
- switching energy density:

$$\frac{C'_{gs} V'_{DD}{}^2}{L'W'} = \frac{\frac{C_{gs} V_{DD}^2}{S^3}}{\frac{LW}{S S}} = \frac{1}{S} \frac{C_{gs} V_{DD}^2}{LW} \downarrow$$

- inverse subthreshold slope:

$$n' = 1 + \frac{C'_{sth}}{C'_{ox}} = 1 + \frac{\sqrt{S} C_{sth}}{S C_{ox}} = 1 + \frac{C_{sth}}{\sqrt{S} C_{ox}} \downarrow$$

but since $V_T \downarrow$, $I_{off} \uparrow\uparrow$.



Two key problems with constant field scaling:

- system designers don't want to scale V_{DD}
- $I_{off} \uparrow\uparrow \Rightarrow$ more static power

- More rigorous study of constant field scaling using 2D simulations
 [P. Vande Voorde, HP Journal, 1997]

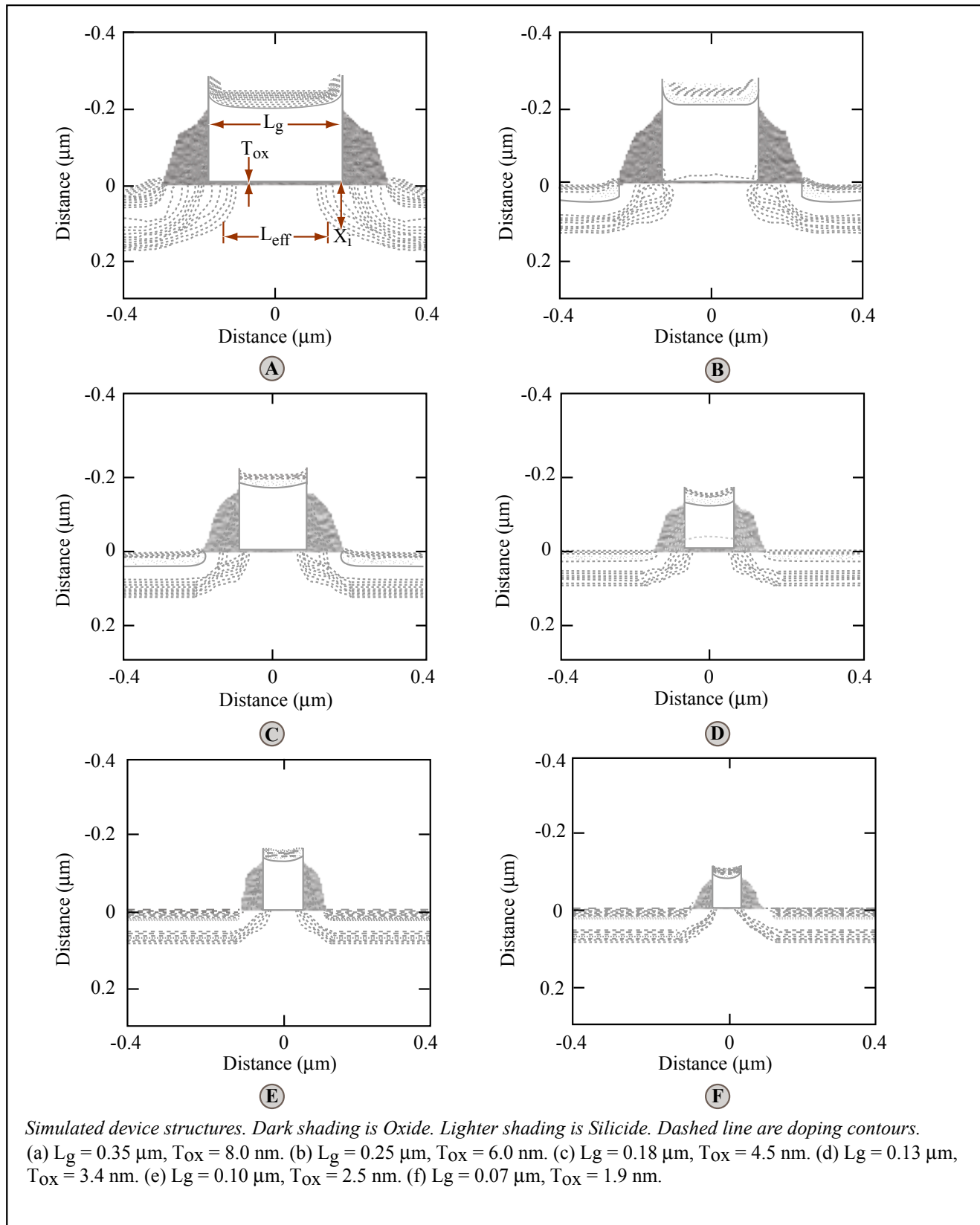


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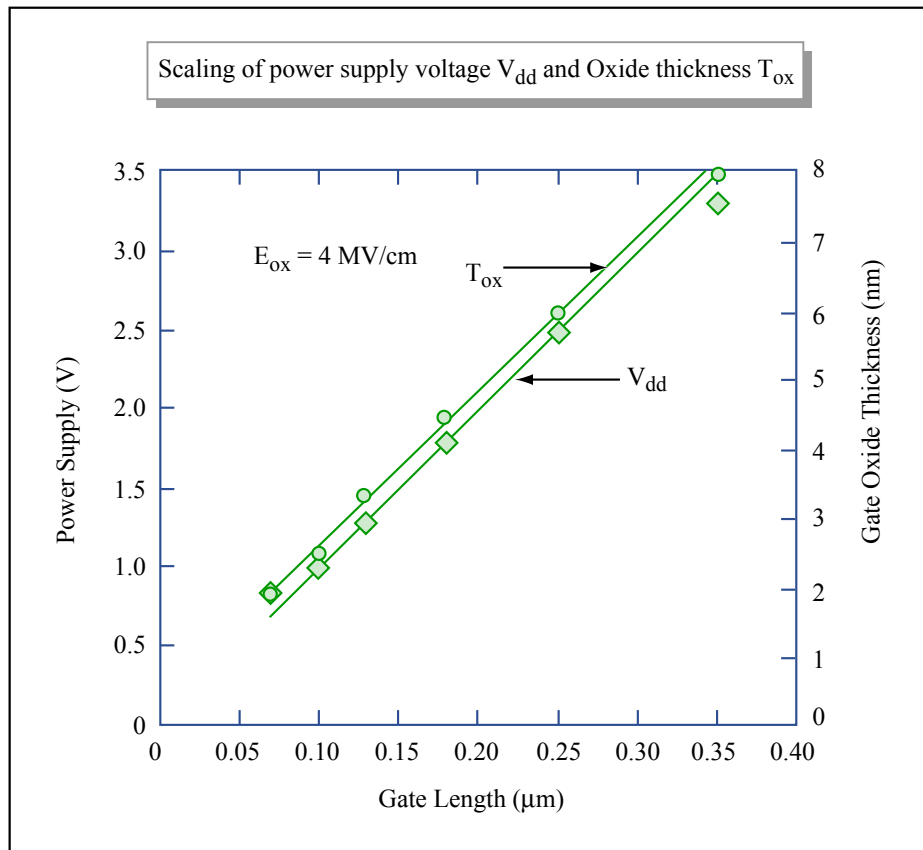


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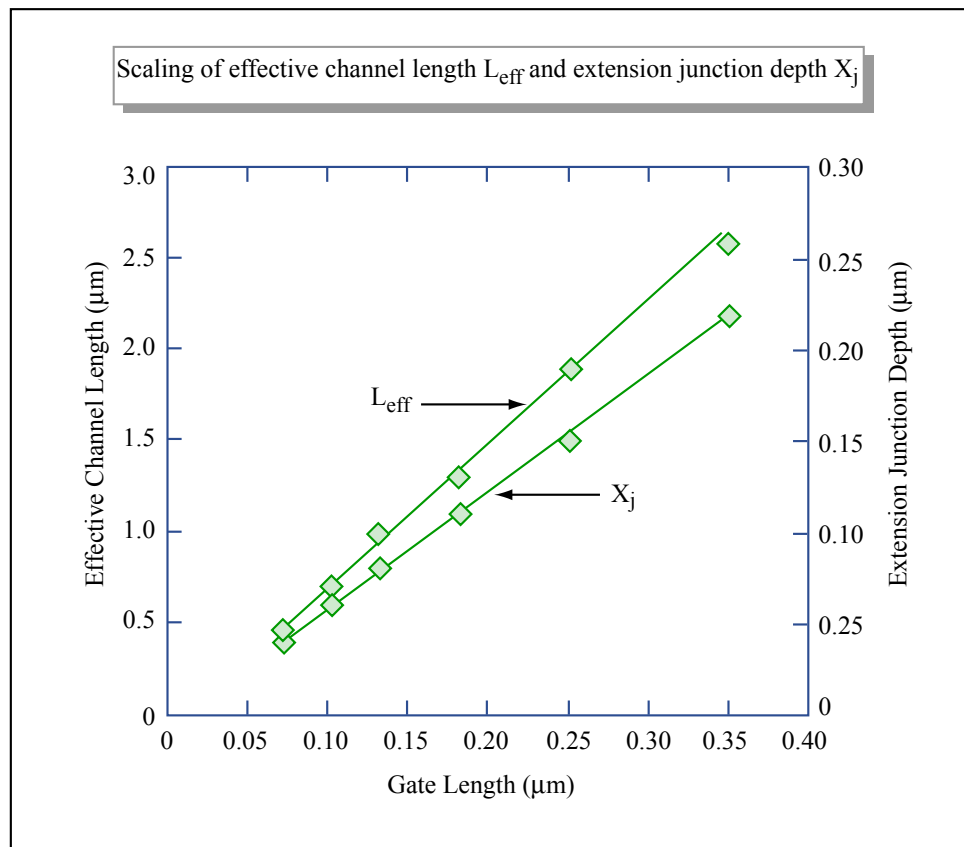


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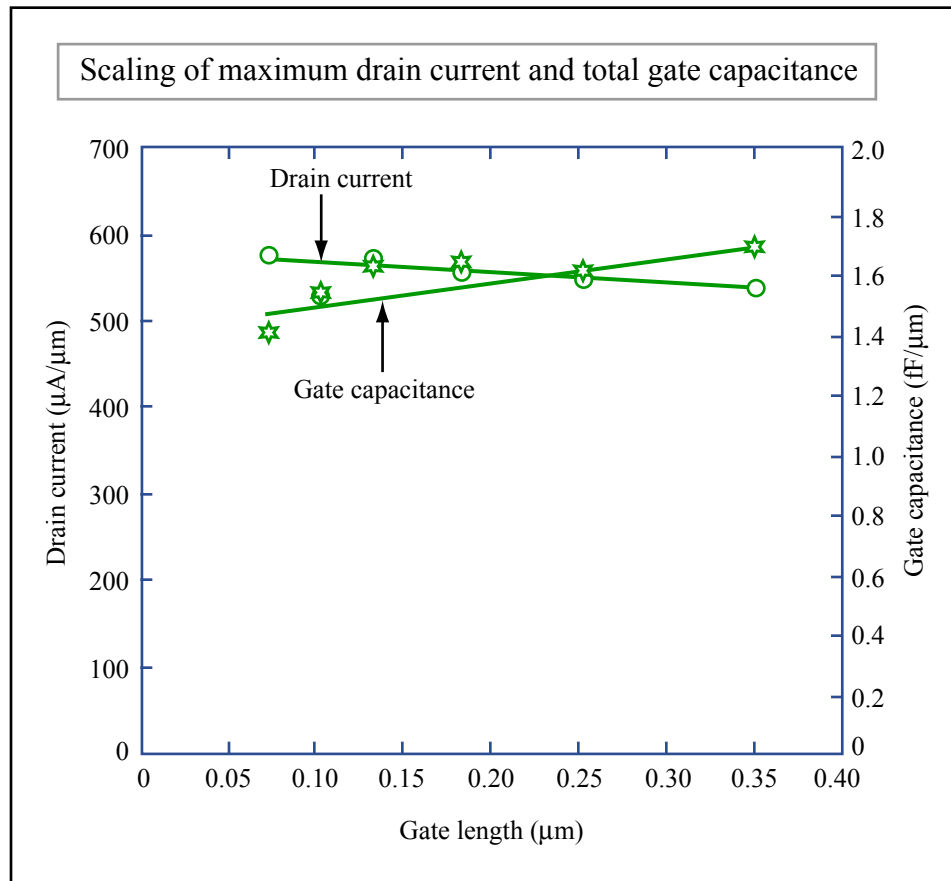


Image by MIT OpenCourseWare. Adapted from Figure 5 on p. 98 in Vande Voorde, Paul. "MOSFET Scaling into the Future." *Hewlett-Packard Journal* (August 1997): 96-100.

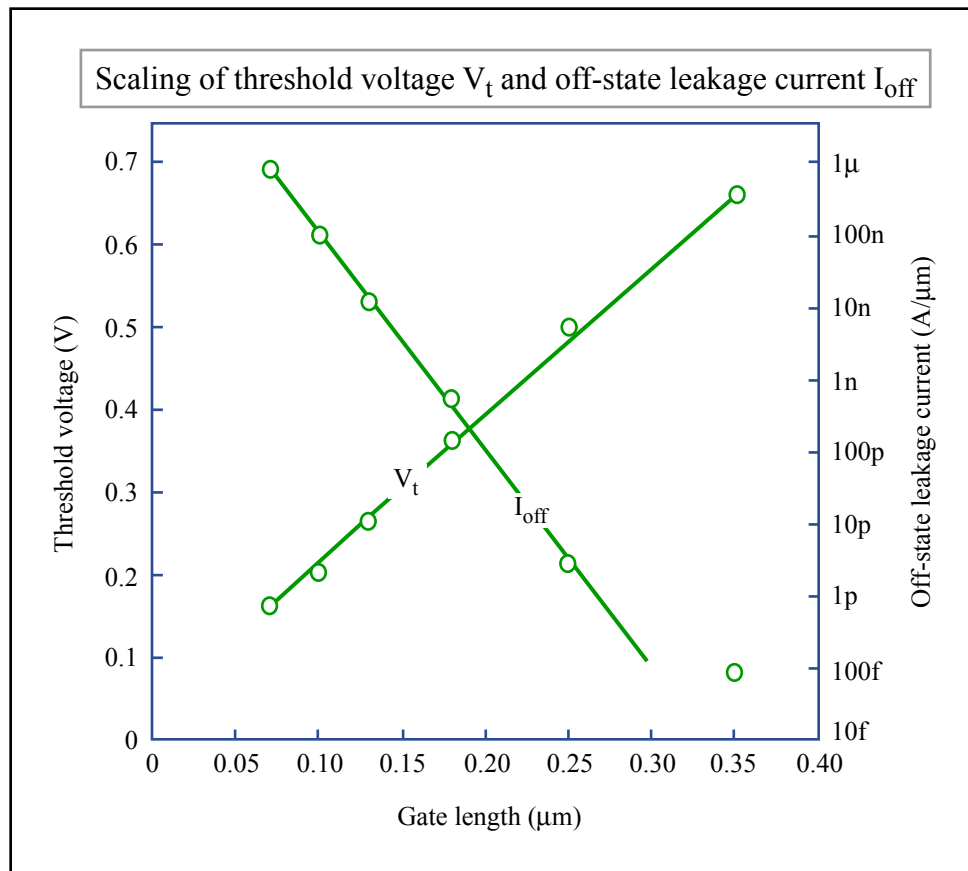


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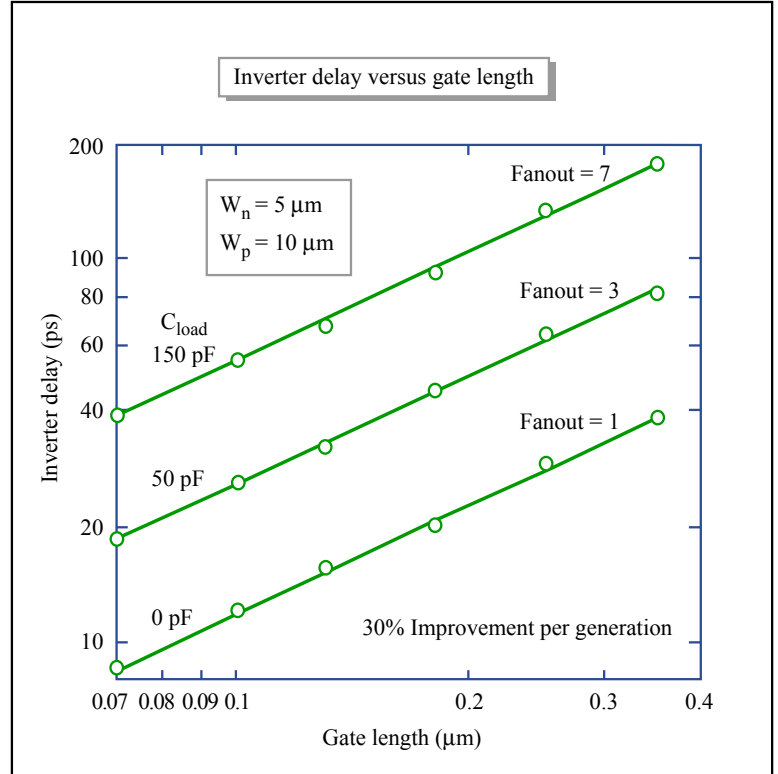
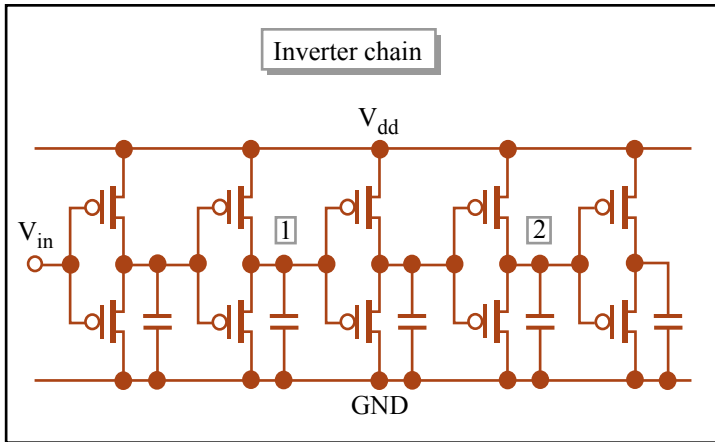


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Image by MIT OpenCourseWare. Adapted from Figure 10 on p. 99 in Vande Voorde, Paul. "MOSFET Scaling into the Future." *Hewlett-Packard Journal* (August 1997): 96-100.

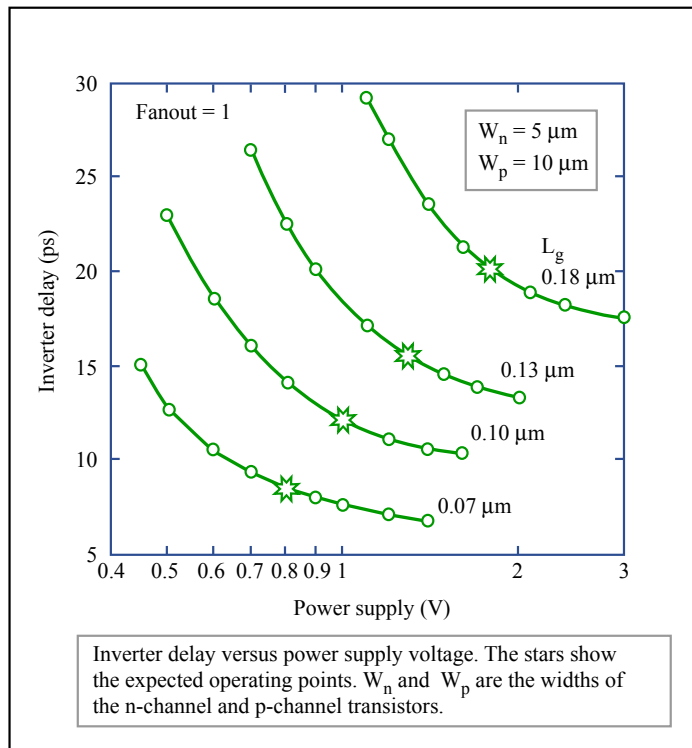


Image by MIT OpenCourseWare. Adapted from Figure 11 on p. 99 in Vande Voorde, Paul. "MOSFET Scaling into the Future." *Hewlett-Packard Journal* (August 1997): 96-100.

□ Constant voltage scaling

Scale all device dimensions but do not scale V_{DD} .

parameter	scaling factor
device dimensions (L, W, x_{ox})	$1/S$
doping level (N_A)	S
supply voltage (V_{DD})	1

Consequences (using long-channel theory):

figure of merit	scaling factor
C_{gs}	$1/S$
V_{th}	$1/\sqrt{S}$
I_D	S
τ	$1/S^2$
$C_{gs}V_{DD}^2$	$1/S$
$C_{gs}V_{DD}^2/LW$	S

Features of constant voltage scaling:

- Performance $\uparrow\uparrow$
- But:
 - It does not address I_{off} problem.
 - Electric field across oxide \uparrow :

$$\mathcal{E}_{ox} = \frac{V_{DD}}{x_{ox}} \propto S \uparrow$$

Reliability problems when $\mathcal{E}_{ox} \simeq 4 \text{ MV/cm}$.

- Electric field in semiconductor (at drain end of channel) \uparrow :

$$\mathcal{E}_m = \sqrt{\frac{(V_{DS} - V_{DSsat})^2}{l^2} + \mathcal{E}_{sat}^2} \propto S \uparrow$$

with

$$l^2 = \frac{\epsilon_s}{\epsilon_{ox}} x_{ox} x_j \propto S^{-2}$$

Reliability problems when $\mathcal{E}_m \simeq 0.5 \text{ MV/cm}$.

- Power density $\uparrow \Rightarrow$ system power \uparrow

□ Generalized scaling

- scale oxide thickness more slowly than other device dimensions
- scale V_{DD} keeping \mathcal{E}_{ox} constant

parameter	scaling factor
L, W	$1/S$
x_{ox}	$1/R$
N_A	S
V_{DD}	$1/R$

with $1 < R < S$.

In generalized scaling:

- I_{off} problem alleviated by not scaling V_T so aggressively;
trade-off: performance
- V_{DD} scales;
trade-off: performance

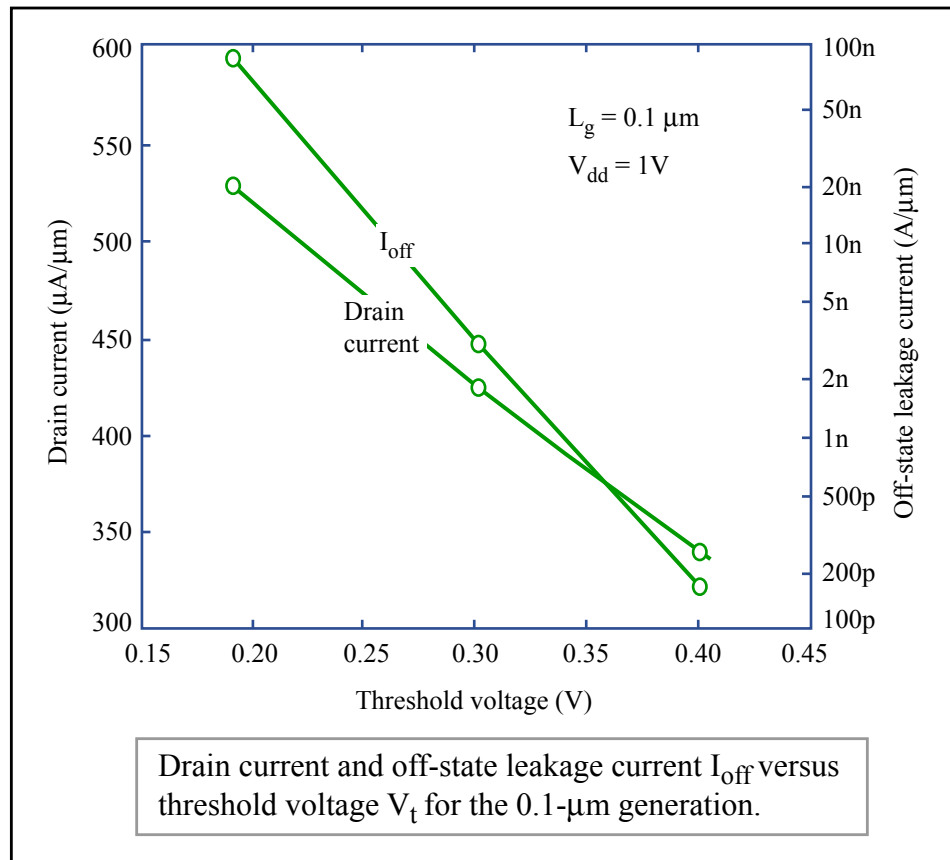


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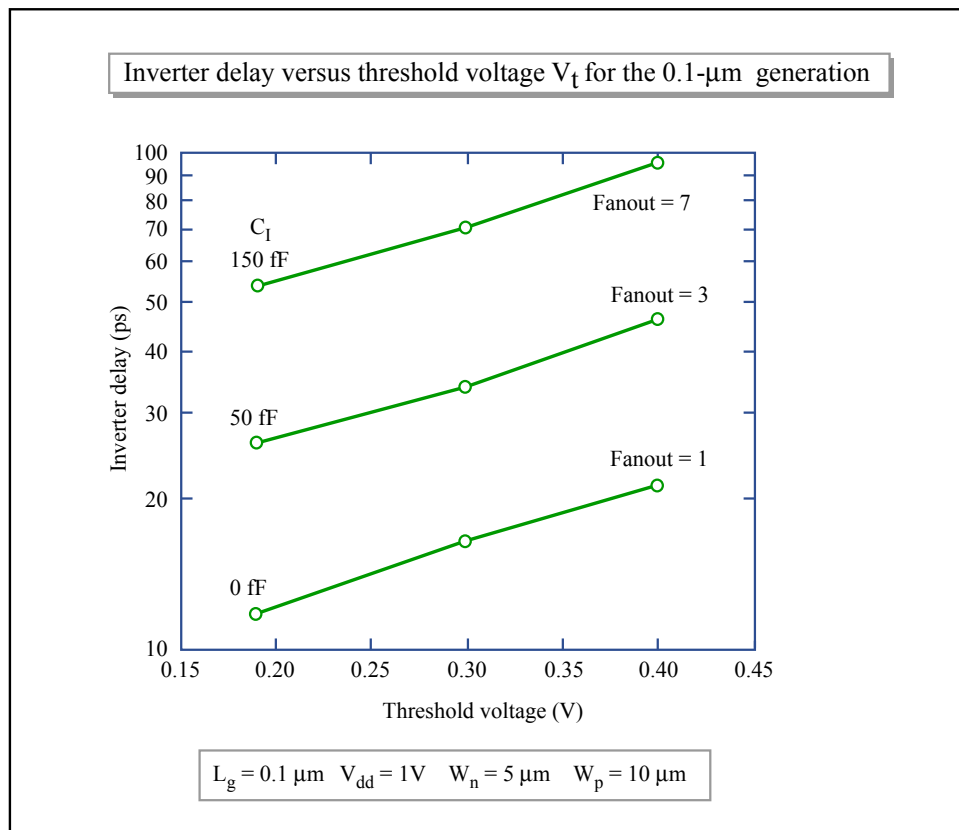
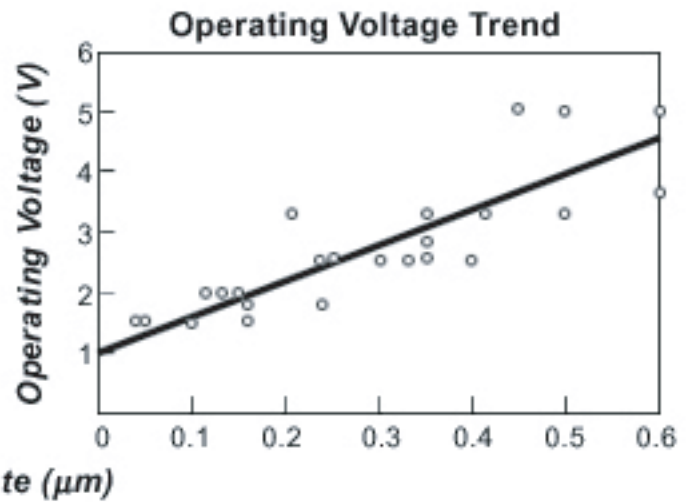
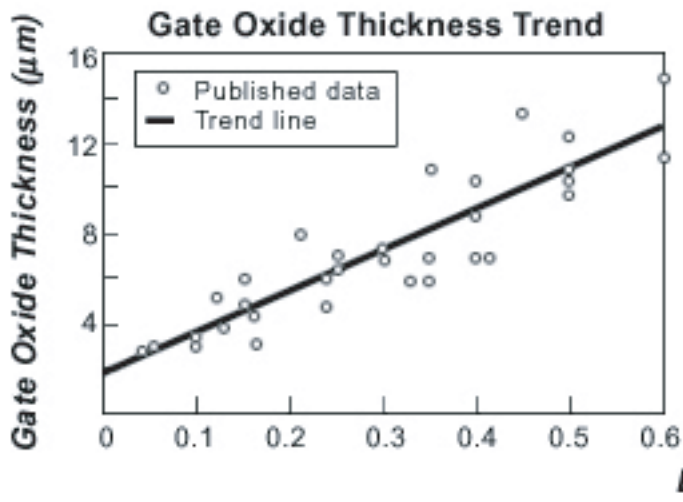
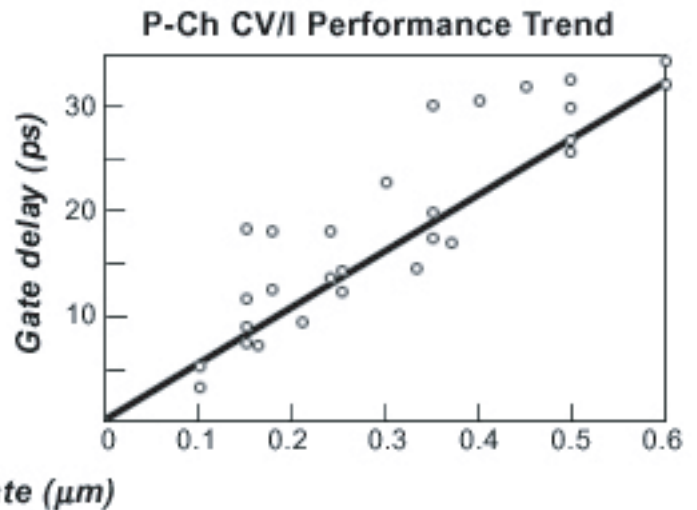
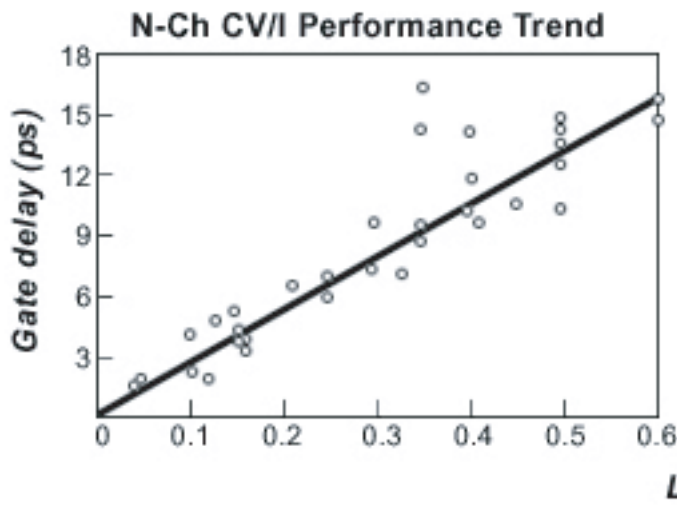


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Oxide thickness will be in the range of 6-10 nm for 0.35 μm generation technologies and will scale to less than 4 nm for 0.10 μm generation technologies.

The 0.35 μm generation marks a transition point between 3.3 V and 2.5 V operating voltage.



N-channel transistor performance trends based on data derived from papers published over the last few years, using the CV/I metric.

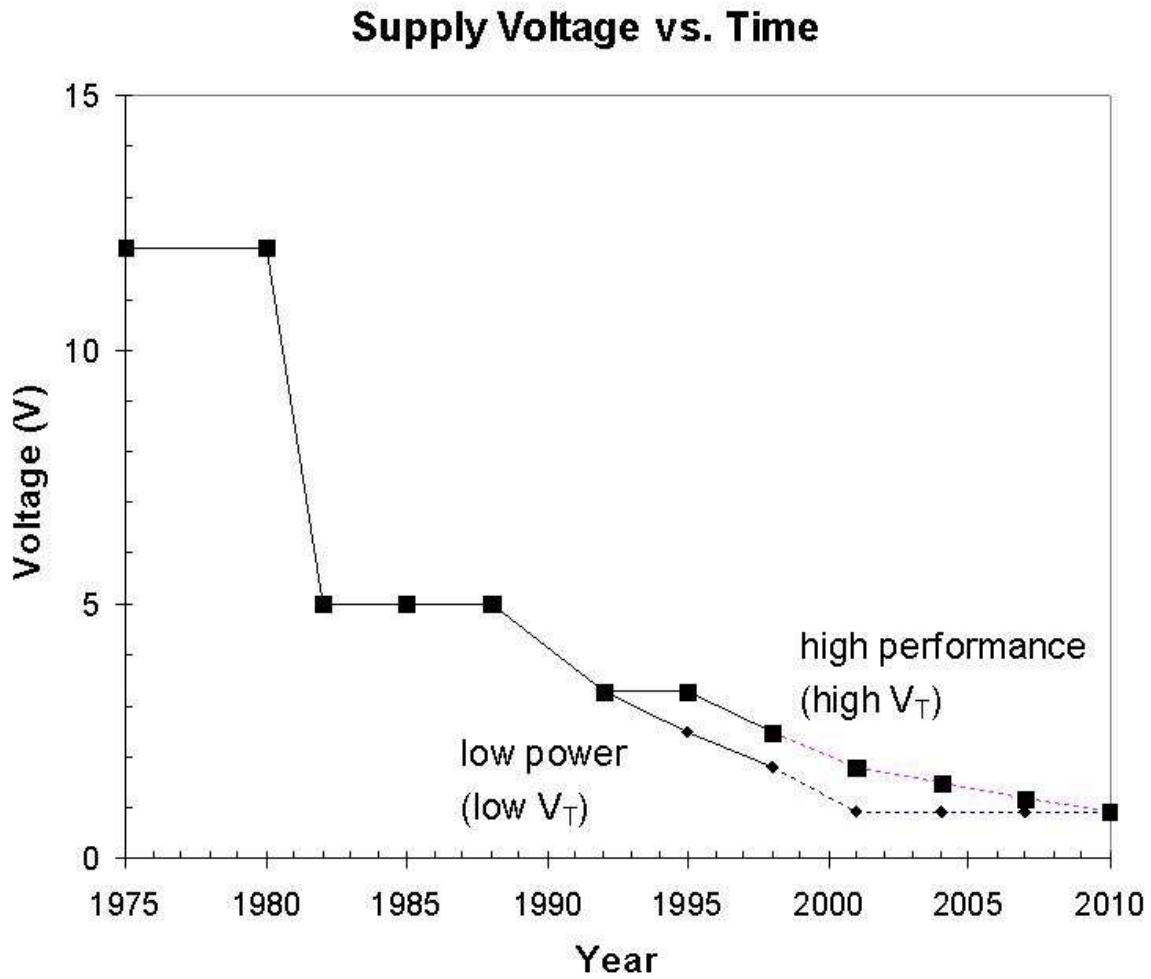
P-channel transistor performance is plotted using the CV/I metric.

Adapted from M. Bohr, *Semiconductor International*, July 1995 (75).

Image by MIT OpenCourseWare. Adapted from Bohr, M. *Semiconductor International* (July 1995): 75.

□ Modern generalized scaling

- Concept of *generation*: every 2 years, new technology is deployed with 30% reduced transistor delay and twice as high transistor density (microprocessor performance doubling every 2 years).
- Everything scales: L (\downarrow), W (\downarrow), x_{ox} (\downarrow), N_A (\uparrow), x_j (\downarrow), and V_{DD} (\downarrow).
- Scaling goal: *extract maximum performance from each generation* (maximize I_{on}), for a given amount of:
 - short-channel effects (DIBL), *and*
 - off-current
- Currently two technology flavors:
 - *high-performance*: high V_{DD} (high I_D , low τ), low V_T (high I_{off});
 - *low-power*: low V_{DD} (low I_D , high τ), high V_T (low I_{off}).



Key conclusions

- *Constant field scaling*: scale all device dimensions keeping vertical and horizontal electric fields constant.

Consequences:

- $I_{off} \uparrow$
- system designers don't want to scale V_{DD}

- *Constant voltage scaling*: scale all device dimensions keeping voltage constant.

Consequences:

- $I_{off} \uparrow$
- fields everywhere $\uparrow \Rightarrow$ reliability compromised

- For a long time scaling proceeded through constant V_{DD} path with abrupt drops in V_{DD} .

- Scaling goal: *extract maximum performance from each generation* (maximize I_{on}), for a given amount of:

- short-channel effects (DIBL), *and*
- off-current

- *Generalized scaling* demands simultaneous scaling of L_g , x_{ox} , x_j , N_A , and V_{DD} .